Lecture 12: Autocovariance as a Pathway to Spectra

Reading: Bendat and Piersol, Ch. 5.2.1

Recap

We've looked at several strategies for computing spectra and estimating degrees of freedom—segmenting data and computing the periodogram (the Fourier transform) or computing the periodogram (Fourier transform) and then averaging. Now, let's look at this from a different angle by considering the covariance.

Using the auto-covariance to think about spectra.

Now let's look at spectra from a different perspective. First, let's remind ourselves of the definition of a convolution:

$$z(t) = \int_{-\infty}^{\infty} x(\tau)y(t-\tau)d\tau. \tag{1}$$

When we talked about Parseval's theorem, my notes mentioned that the autocovariance is the convolution of x(t) with its time reversal, x(-t).

$$y(\tau) = \int_{-\infty}^{\infty} x(t)x(\tau + t)dt.$$
 (2)

More formally, we might write this autocovariance as $R_{xx}(\tau)$.

$$R_{xx}(\tau) = \int_{-\infty}^{\infty} x(t)x(\tau+t)dt.$$
 (3)

Now, what if we Fourier transform R?

$$S_{xx}(f) = \int_{-\infty}^{\infty} R_{xx}(\tau)e^{-i2\pi f\tau} d\tau. \tag{4}$$

Formally, this and its inverse transform are the Wiener-Khinchine relations.

Now let's think about starting with two functions, x(t) and y(t). We can write their Fourier transforms:

$$X(f) = \int_{-\infty}^{\infty} x(t)e^{-i2\pi ft} dt$$
 (5)

$$Y(f) = \int_{-\infty}^{\infty} y(t)e^{-i2\pi ft} dt.$$
 (6)

So now let's define X times the complex conjugate of Y. (Why do we consider the complex conjugate? Because it's how we always multiply vectors.) We find the Fourier transform of the complex conjugate by substituting -i for i everywhere it appears:

$$Y^*(f) = \int_{t=-\infty}^{\infty} y(t)e^{i2\pi ft} dt \tag{7}$$

$$= \int_{-t=-\infty}^{\infty} y(-t)e^{-i2\pi ft} d(-t)$$
 (8)

$$= \int_{t=\infty}^{-\infty} -y(-t)e^{-i2\pi ft} dt$$
 (9)

$$= \int_{t=-\infty}^{\infty} y(-t)e^{-i2\pi ft} dt. \tag{10}$$

So

$$X(f)Y^*(f) = \int_{-\infty}^{\infty} k(t)e^{-i2\pi ft} dt.$$
(11)

For the moment, we have no idea what k(t) should be, but we should be able to figure it out. If $X(f)Y^*(f)$ is a product in the frequency domain, then k(t) should be a convolution in the time domain:

 $k(t) = \int_{-\infty}^{\infty} x(u)y(u-t) du.$ (12)

(Remember that we'd normally use t-u for a convolution; but here we reverse the sign to be consistent with using the complex conjugate $Y^*(f)$.) We used a derivation very similar to this in about Lecture 7, when we wanted to persuade ourselves that Parseval's theorem would work. But now we revisit this with a goal of looking closely at this convolved quantity, which represents the autocovariance. We can plug k(t) into our equation to check this.

$$\int_{-\infty}^{\infty} k(t)e^{-i2\pi ft} dt = \int_{-\infty}^{\infty} \left\{ \int_{-\infty}^{\infty} x(u)y(u-t) du \right\} e^{-i2\pi ft} dt$$
 (13)

$$= \int_{-\infty}^{\infty} \int_{u-t=\infty}^{-\infty} -x(u)y(u-t)e^{i2\pi f(u-t)}e^{-i2\pi f u} d(u-t) du$$
 (14)

$$= \int_{-\infty}^{\infty} x(u)e^{-i2\pi f u} \left\{ \int_{-\infty}^{\infty} y(u-t)e^{i2\pi f(u-t)} d(u-t) \right\} du \qquad (15)$$

$$= \int_{-\infty}^{\infty} x(u)e^{-i2\pi f u}Y^*(f) du \tag{16}$$

$$= Y^*(f) \int_{-\infty}^{\infty} x(u)e^{-i2\pi f u} du$$
 (17)

$$= X(f)Y^*(f). (18)$$

Here we've taken advantage of the fact that the integral runs from $-\infty$ to $+\infty$ which lets us treat u-t as a variable that depends only on t.

So we can think about what happens when x(t) = y(t), so that

$$k(t) = \int_{-\infty}^{\infty} x(u)x(u-t) \, du = R_{xx}(-t). \tag{19}$$

This means that k(t) is the autocovariance of x. The autocovariance is symmetric, so we could also write this as

$$k(t) = \int_{-\infty}^{\infty} x(u+t)x(u) \, du = R_{xx}(t). \tag{20}$$

Regardless

$$|X(f)|^2 = \int_{-\infty}^{\infty} k(t)e^{-i2\pi ft} dt.$$
 (21)

This says that the Fourier transform coefficients squared (what we use when we compute spectra) are equivalent to the Fourier transform of the autocovariance.

A practical look at the autocovariance

Let's think about the autocovariance of white noise:

$$R_{xx}(\tau) = \int_{-\infty}^{\infty} x(t)x(\tau+t)dt = \begin{cases} 0 & \text{for } \tau \neq 0\\ 1 & \text{for } \tau = 0 \end{cases}$$
 (22)

This is true, because white noise is uncorrelated except at zero lag.

Alternatively, if I consider red noise, then the noise will be correlated from point to point, and the autocovariance will have a bit of spread. We can test this out:

```
% define red data with autoregressive process
a=randn(10000,1);
b(1)=a(1);
for i=2:length(a)
  b(i)=b(i-1)+a(i);
end

BB=xcorr(b,b)/max(xcorr(b,b));
BB_unbiased=xcorr(b,b,'unbiased')/max(xcorr(b,b,'unbiased'));
plot(-999:999,[BB' BB_unbiased'],'LineWidth',3)
```

Biased vs unbiased estimators.

Notice that we could consider both the 'biased' and the "unbiased" estimator. There are arguments for either choice. The difference depends on how we normalize our discrete autocovariance. In the unbiased case, we're computing

$$R(\tau)_{unbiased} = \frac{1}{N-m} \sum_{n=1}^{N-m} x(t_n) x(t_{n+m}).$$
 (23)

In the biased case, we change how we normalize:

$$R(\tau)_{biased} = \frac{1}{N} \sum_{n=1}^{N} x(t_n) x(t_{n+m}),$$
 (24)

which means that as the number of values we consider becomes smaller, we constrain the magnitude of the autocovariance by continuing to divide by N. Emery and Thomson note that the biased estimator acts like a triangle window. For spectra, you will want to stick to the unbiased estimator, and then choose your own window carefully.

Practical implementation: Spectra from auto-covariance

Using the auto-covariance to compute spectra requires averaging, just as we did by segmenting our data and using the fft, but there's one tidy little trick. Let's use some white noise again, and take a look at our options:

1. Suppose we start with a big matrix of white noise, and we compute the autocovariance for each column of our matrix, then Fourier transform, and use these to compute a spectrum. We'll end up doing something along these lines:

```
A=randn(1000,100);
for i=1:100
  AcA(:,i)=xcov(A(:,i),A(:,i),'unbiased'); % autocovariance for end
fAcA=fft(AcA(500:1500,:)); % Fourier transform of autocovariance
```

```
frequency=(0:500)/1000;
loglog(frequency,2*abs(mean(fAcA(1:501,:),2)),'LineWidth',3)
set(gca,'FontSize',16)
xlabel('Frequency (cycles per data point)','FontSize',16)
ylabel('Spectral energy','FontSize',16)
```

As with the standard Welch method for spectral estimation (which we can properly refer to as a "periodogram"), we're only plotting half the spectrum, so we need double the energy.

2. Alternatively, we could average all of the autocovariances, and then Fourier transform:

```
mean_AcA=mean(AcA,2);
fmean_AcA=fft(mean_AcA(500:1500));
hold on
loglog(frequency,2*abs(fmean_AcA(1:501,:))*1.1,'r','LineWidth',3)
legend('average of FFTs of many autocovariances',...
    'FFT of averaged autocovariance (scaled by 1.1)')
```

3. For comparison, the peridogram-based determined from the fft of the data:

```
fA=fft(A);
ampA=abs(fA(1:501,:)).^2/1000; ampA(2:500,:)=2*ampA(2:500,:);
loglog(frequency, mean(ampA, 2), 'LineWidth', 3)
```

4. Finally, you might skip all segmenting and just compute the autocovariance of the full record:

```
N=length(a);
aca=xcov(a,a,'unbiased');
faca=fft(aca(N-500:N+500));
loglog(frequency,2*abs(faca(1:501)),'LineWidth',3)
```

In the results, shown in Figure 1, the curves (for options 1 and 2) are identical, though the red line has been scaled up by 10% to make both visible. There are some normalizations here that we haven't properly confronted. (Notably, if our frequency had units, we'd need to scale by T/N, or include a Δt .) Details can be sorted out later, and Thomson and Emery provide a bit of guidance on this.

You shouldn't be surprised that averaging before or after the FFT leads to the same results, since averaging has no impact on the FFT. This might give you ideas of how you can take advantage of the autocovariance to compute spectra from gappy data. Importantly, we can compute spectra without needing to chunk our data and compute lots of ffts, provided that we had a good estimate of the autocovariance.

The best estimate of the autocovariance should have the most averaging, and you can maximize your averages by using the longest possible records (option 4 above). Suppose you have 1000 data points, and two possible ways to analyze the data.

• If you split the record into 10 segments of 100 points, the zero-lag autocovariance will be an average of 1000 points, which is great, but the lag 100 autocovariance will represent an average of only 10 points, which is not so great.

• In contrast, if you don't split the record, the zero-lag autocovariance will still be an average of 1000 points, while the lag 100 auto-covariance will be an average of 900 points, on the whole a lot more averaging.

In the days before the development of the FFT, the autocovariance was a natural pathway to determining the spectrum, since it was clean and easy to compute. And now, with modern computing, you might not feel like there's any need to take advantage of the FFT anymore. If you can obtain the best possible estimate of the autocovariance, by whatever means necessary, then you should be able to compute one FFT and obtain reasonable estimate of the spectrum, without concern for data gaps or computational speed.

Formally if you compute spectra from the autocovariance, you need to think about averaging just as thoroughly as you do for segmented windowed data (through the Welch method) or for frequency-averaged spectra (the Daniell method). In this case, the challenge comes in deciding what fraction of the autocovariance to actually Fourier transform. If you use the autocovariance over the entire data range, your autocovariance estimator has too much uncertainty at the large lags, and the resulting spectrum will have large uncertainties.

When we use the autocovariance to compute spectra, we'll want to omit the poorly sampled edges of the spectrum. There are a number of ways to do this. We have to decide what fraction of the autocovariance to use. For example I could take half, or a quarter. What difference does it make? In essence, the fraction that I use determines the amount of averaging that I do, and therefore defines the number of degrees of freedom. In the simplest form, we use a boxcar window to extract values:

The degrees of freedom are determined by the width of the window. If I have N data points, and I use M of them in the Fourier transform, then I'm implicitly averaging N/M independent samples, which gives me N/M degrees of freedom. I can compute an error bar based on this in exactly the way that we did earlier for the periodogram approach:

```
hold on
nu=N/M;
err_high=nu/chi2inv(.05/2,nu);
err_low=nu/chi2inv(1-.05/2,nu);
loglog([10 10],[err_high err_low],'LineWidth',3)
```

Now you might decide that a boxcar windowed view of the autocovariance is likely to have unfortunate characteristics in the Fourier domain, in which case you could use a different window instead: a triangle window (also called a Bartlett window), or a Hanning window, or a Hamming window. For example:

```
fBB t=fft(BB(N-M:N+M).*triang(2*M+1)');
```

```
fBB_h=fft (BB (N-M:N+M) .*hanning (2*M+1)');
loglog(0:M-1,abs(fBB_t(1:M)))
loglog(0:M-1,abs(fBB_h(1:M)))
```

In this case we adjust the error bars to account for the windowing of the autocovariance. As it turns out, the table (5.5 in Thomson and Emery, 2014) that was so misleading for overlapping segmented data is exactly what we need here:

Window type	Equivalent degrees
	of freedom (ν)
Truncated peridogram (boxcar)	N/M
Bartlett (triangle)	3N/M
Daniell (sinc)	2N/M
Parzen	3.708614N/M
Hanning	8/3N/M
Hamming	2.5164N/M

This works well when M is small compared with N and when the autocovariance is comparatively narrow. In other words, for white noise, this converges nicely; for a broad red noise peak, the autocovariance tends to have negative lobes, making the choice of M difficult and the results not amenable to interpretation. The challenge in dealing with the truncation of the autocovariance is perhaps the reason that although this approach is often presented in equations (in idealized cases with infinite data), it is less frequently implemented for real applications.

Side bar: Autocovariance in discrete form.

Last time we went through the representation of the autocovariance using an integral form. Let's rewrite this in terms of the discrete Fourier transform. In this case, the mean of our data is:

$$\langle x \rangle = \frac{1}{2T} \int_{-T}^{T} x(t)e^{i0} dt = a_0.$$
 (25)

and the variance is

$$\langle x * x \rangle = \frac{1}{2T} \int_{-T}^{T} x^*(t) x(t) dt - |a_0|^2.$$
 (26)

We use the complex conjugate here, just in case x(t) is represented as a complex number, since this will give us the sum of the squares. Notice that we've remembered to subtract out the mean (our frequency zero Fourier coefficient).

In similar notation, we can write the covariance (for finite record length 2T) as:

$$R(\tau) = \frac{1}{2T} \int_{-T}^{T} x^*(t) x(t+\tau) dt - |a_0|^2.$$
 (27)

This lets us write out an expression for the variance R in terms of the discrete Fourier coefficients:

$$R(\tau) = \frac{1}{2T} \int_{-T}^{T} \left[\sum_{n=-\infty}^{\infty} a_n^* e^{-i2\pi f_n t} \sum_{m=-\infty}^{\infty} a_m e^{i2\pi f_m (t+\tau)} \right] dt - |a_0|^2$$
 (28)

$$= \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} a_n^* a_m e^{+i2\pi f_m \tau} \frac{1}{2T} \int_{-T}^{T} e^{i(-2\pi f_n + 2\pi f_m)t} dt - |a_0|^2$$
 (29)

$$= \sum_{m=-\infty}^{\infty} |a_m|^2 e^{+i2\pi f_m \tau} - |a_0|^2 \tag{30}$$

where we used a Kronecker delta (δ_{nm}) to eliminate the integral with $e^{2\pi i(-f_n+f_m)t}$ except when n=m, and we subtracted a_0^2 at the end to match our original definition. In setting this up, recall (from lecture 5) that the Fourier transform uses $e^{-i2\pi ft}$, and the inverse transform uses $e^{+i2\pi ft}$. We're using the inverse transform here (though the signs reversed when we had the complex conjugate. This tells us that the Fourier transform of the autocovariance can be expressed by the squared Fourier coefficients. (So we could avoid the Fourier transform completely and just work with the auto-covariance.)

In this form, Parseval's theorem simply says that

$$R(0) = \frac{1}{2T} \int_{-T}^{T} x^*(t)x(t+0) dt - |a_0|^2$$
(31)

$$= \sum_{m=-\infty}^{\infty} |a_m|^2 - |a_0|^2 \tag{32}$$

meaning that the variance of x is the sum of magnitudes of the Fourier coefficients.

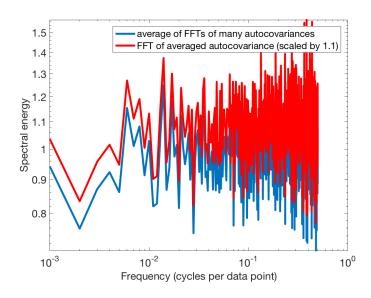


Figure 1: Spectra for white noise, computed by Fourier transforming 100 realizations of the autocovariance function (blue), or by Fourier transforming a smoothed autocovariance function computed from 100 realizations of the data (red). The red line is scaled upward by a factor of 1.1.